Stochastic Adaptive Control (02421)

Lecture 11

DTU Compute

Tobias K. S. Ritschel

Section for Dynamical Systems

Department of Applied Mathematics and Computer Science

Technical University of Denmark

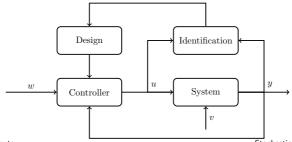
 $f(x+\Delta x) = \sum_{i=0}^{\infty} \frac{(\Delta x)^i}{i!} f^{(i)}(x)$ Department of Applied Mathematics and Computer Science

Stochastic Adaptive Control - Follow-up Lecture Plan

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- 1 System theory
- 2 Stochastics
- **3** State estimation 1
- 4 State estimation 2
- **6** Optimal control 1
- **6** System identification 1 + adaptive control 1
- **7** External models + prediction

- Optimal control 2
- 9 Optimal control 3
- Ø System identification 2
- System identification 3 + model validation
- System identification 4 + adaptive control 2
- Adaptive control 3



Stochastic Adaptive Control - Follow-up Today's Agenda

- Time-varying estimation
- Design of experiments
- Model validation

Stochastic Adaptive Control - Follow-up Follow-up from last time



Questions?

Heuristics for time-varying systems

Stochastic Adaptive Control - Time-variant Estimation Time-varying estimation - first example

ARX model

$$A(q^{-1})y_t = B(t, q^{-1})u_t + e_t,$$
(1)

$$b_1(t) = b_{1,0} + b_{1,1}t \tag{2}$$

Treat time-varying coefficient as two coefficients with their own inputs

$$y_t = \phi^T \theta + e_t \tag{3}$$

$$\theta^T = \begin{bmatrix} a_1 & a_2 & \cdots & a_{n_a} & b_{1,0} & b_{1,1} & b_2 & \cdots & b_{n_b} \end{bmatrix}$$
(4)

$$\phi^{T} = \begin{bmatrix} -y_{t-1} & -y_{t-2} & \cdots & -y_{t-n_a} & u_{t-1} & tu_{t-1} & u_{t-2} & \cdots & u_{t-n_b} \end{bmatrix}$$
(5)

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Stochastic Adaptive Control - Time-variant Estimation Time-varying estimation

For deterministic time varying systems, rearrange the parameters

$$y_t = \phi_t^T \theta_t + e_t \tag{6}$$

$$\theta_t = \alpha + f(t)\beta = \begin{bmatrix} I & f(t) \end{bmatrix} \begin{bmatrix} \alpha \\ \beta \end{bmatrix}$$
(7)

$$y_t = \begin{bmatrix} \phi_t^T & \phi_t^T f(t) \end{bmatrix} \begin{bmatrix} \alpha \\ \beta \end{bmatrix} + e_t$$
(8)

For piece-wise linear parameters, rearrange the parameters

$$y_t = \phi_t^T \theta_t + e_t \tag{9}$$

$$\theta_t = \alpha_i + (t - T_i)\beta_i, \quad T_i \le t \le T_{i+1}$$
(10)

$$y_t = \begin{bmatrix} \phi_t^T & \phi_t^T (t - T_i) \end{bmatrix} \begin{bmatrix} \alpha_i \\ \beta_i \end{bmatrix} + e_t$$
(11)

Stochastic Adaptive Control - Time-variant Estimation Time-varying systems

System with general time-varying parameters

$$\theta_{t+1} = f(t, \theta_t, v_t) \tag{12}$$

- The methods discussed so far cannot estimate the time-varying dynamics and were not designed to do it
- In practice, the problem is that the correction factor diminishes over time

$$P_t \to 0 \tag{13}$$



Reset the covariance after some time, t_i

$$P_{t_i} = P_i > P_{t_i-1}, \quad \hat{\theta}_{t_i} = \hat{\theta}_{t_i-1}$$
 (14)

The appropriate restarting time depends on the application

For instance, restart at fixed intervals

$$t_i = Ni \tag{15}$$

This can be useful for periodic systems

Stochastic Adaptive Control - Time-variant Estimation DTU Time-varying systems - Forgetting methods: Constant Gain

Another method: Keep the correction term large

For instance, keep the correction term κ constant

$$\hat{\theta}_t = \hat{\theta}_{t-1} + \kappa \epsilon_t \tag{16}$$

$$\tilde{\theta}_t = (I - \kappa \phi_t^T) \tilde{\theta}_t - \kappa e_t \tag{17}$$

Alternatively, keep the variance constant

$$P_t = P \tag{18}$$

$$\hat{\theta}_t = \hat{\theta}_{t-1} + \kappa \epsilon_t \tag{19}$$

$$\kappa_t = \frac{P\phi_t}{1 + \phi_t^T P\phi_t} \tag{20}$$

Stochastic Adaptive Control - Time-variant Estimation Time-varying systems - Forgetting methods: Exponential Forgetfulness

Another approach: Forget a little bit all the time (exponential forgetfulness)

$$J_{t} = \frac{1}{2} \sum_{i=1}^{t} \lambda^{t-i} \epsilon_{i}^{2} = \lambda J_{t-1} + \frac{1}{2} \epsilon_{t}^{2}$$
(21)

The recursion is similar to the previous methods

$$\hat{\theta}_t = \hat{\theta}_{t-1} + P_t \phi_t \epsilon_t \tag{22}$$

$$\epsilon_t = y_t - \phi_t^T \hat{\theta}_{t-1} \tag{23}$$

$$P_t^{-1} = \lambda P_{t-1}^{-1} + \phi_t \phi_t^T$$
(24)

The forgetting factor λ can be expressed in terms of a horizon, N_∞

$$\lambda = 1 - \frac{1}{N_{\infty}} \tag{25}$$

18.4.2024

Model

$$y_t = \frac{-1}{4}y_{t-1} + \frac{1}{2}y_{t-2} + u_{t-1} + e_t$$
(26)

New measurement (t = 10)

$$y_t = 1.47, u_{t-1} = 2$$
 (27)

Covariates, parameter estimate and covariance

$$\phi_t = \begin{bmatrix} 2.42\\ 2.57\\ 2 \end{bmatrix}, \quad \hat{\theta}_{t-1} = \begin{bmatrix} -0.2505\\ 0.4960\\ 0.9991 \end{bmatrix}, \quad P_{t-1} = \begin{bmatrix} 0.1355 & -0.0431 & -0.1057\\ -0.0431 & 0.0620 & -0.0088\\ -0.1057 & -0.0088 & 0.1242 \end{bmatrix}$$

Forgetting factor

$$\lambda = 0.95 \tag{28}$$

Stochastic Adaptive Control - Time-variant Estimation Exponential forgetting - Example

Residual

$$\epsilon_t = y_t - \phi_t^T \hat{\theta}_{t-1} = 1.47 - \begin{bmatrix} 2.42 & 2.57 & 2 \end{bmatrix} \begin{bmatrix} -0.2505\\ 0.4960\\ 0.9991 \end{bmatrix} = -1.1967 \quad (29)$$

Covariance

$$P_{t}^{-1} = \lambda P_{t-1}^{-1} + \phi_{t} \phi_{t}^{T} = 0.95 \begin{bmatrix} 259.9763 & 214.2840 & 236.4348 \\ 214.2840 & 192.9152 & 196.0344 \\ 236.4348 & 196.0344 & 223.1583 \end{bmatrix}$$
(30)
+
$$\begin{bmatrix} 2.42 \\ 2.57 \\ 2 \end{bmatrix} \begin{bmatrix} 2.42 & 2.57 & 2 \end{bmatrix} = \begin{bmatrix} 252.8339 & 209.7892 & 229.4530 \\ 209.7892 & 189.8743 & 191.3727 \\ 229.4530 & 191.3727 & 216.0004 \end{bmatrix}$$
(31)



$$\hat{\theta}_{t} = \hat{\theta}_{t-1} + P_{t}\phi_{t}\epsilon_{t} = \begin{bmatrix} -0.2505\\ 0.4960\\ 0.9991 \end{bmatrix}$$

$$-1.1967 \begin{bmatrix} 0.1426 & -0.0456 & -0.1111\\ -0.0456 & 0.0638 & -0.0081\\ -0.1111 & -0.0081 & 0.1298 \end{bmatrix} \begin{bmatrix} 2.42\\ 2.57\\ 2 \end{bmatrix} = \begin{bmatrix} -0.2574\\ 0.4512\\ 1.0350 \end{bmatrix}$$
(33)

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Model

$$y_t = \frac{-1}{4}y_{t-1} + \frac{1}{2}y_{t-2} + u_{t-1} + e_t$$
(34)

New measurement (t = 10)

$$y_t = 1.47, u_{t-1} = 2$$
 (35)

Covariates, parameter estimate and covariance

$$\phi_t = \begin{bmatrix} 2.42\\ 2.57\\ 2 \end{bmatrix}, \quad \hat{\theta}_{t-1} = \begin{bmatrix} -0.2505\\ 0.4960\\ 0.9991 \end{bmatrix}, \quad P_{t-1} = \begin{bmatrix} 0.1355 & -0.0431 & -0.1057\\ -0.0431 & 0.0620 & -0.0088\\ -0.1057 & -0.0088 & 0.1242 \end{bmatrix}$$

Forgetting factor

$$\lambda = 0.90 \tag{36}$$

Solve the exercise in 15 min.

15 DTU Compute

Stochastic Adaptive Control - Time-variant Estimation Exponential forgetting - Exercise

Residual

$$\epsilon_t = y_t - \phi_t^T \hat{\theta}_{t-1} = 1.47 - \begin{bmatrix} 2.42 & 2.57 & 2 \end{bmatrix} \begin{bmatrix} -0.2505\\ 0.4960\\ 0.9991 \end{bmatrix} = -1.1967 \quad (37)$$

Covariance

$$P_{t}^{-1} = \lambda P_{t-1}^{-1} + \phi_{t} \phi_{t}^{T} = 0.90 \begin{bmatrix} 259.9763 & 214.2840 & 236.4348 \\ 214.2840 & 192.9152 & 196.0344 \\ 236.4348 & 196.0344 & 223.1583 \end{bmatrix}$$
(38)
+
$$\begin{bmatrix} 2.42 \\ 2.57 \\ 2 \end{bmatrix} \begin{bmatrix} 2.42 & 2.57 & 2 \end{bmatrix} = \begin{bmatrix} 239.8351 & 199.0750 & 217.6313 \\ 199.0750 & 180.2286 & 181.5709 \\ 217.6313 & 181.5709 & 204.8424 \end{bmatrix}$$
(39)



Parameter estimate

$$\hat{\theta}_{t} = \hat{\theta}_{t-1} + P_{t}\phi_{t}\epsilon_{t} = \begin{bmatrix} -0.2505\\ 0.4960\\ 0.9991 \end{bmatrix}$$

$$-1.1967 \begin{bmatrix} 0.1505 & -0.0481 & -0.1172\\ -0.0481 & 0.0672 & -0.0085\\ -0.1172 & -0.0085 & 0.1369 \end{bmatrix} \begin{bmatrix} 2.42\\ 2.57\\ 2 \end{bmatrix} = \begin{bmatrix} -0.2577\\ 0.4488\\ 1.0369 \end{bmatrix}$$
(41)

Stochastic Adaptive Control - Time-variant Estimation

Time-varying systems - Fortescue's Method



Improve with a time-varying forgetting factor depending on the prediction error, $\boldsymbol{\epsilon}_t$

$$\lambda_t = 1 - \frac{1}{N_0} \frac{\epsilon_t^2}{\sigma^2 s_t} \tag{42}$$

 ${\it N}_0$ is the approx. horizon over which the parameter is roughly constant

Recursion

$$\epsilon_t = y_t - \phi_t^T \hat{\theta}_{t-1} \tag{43}$$

$$s_t = 1 + \phi_t^T P_{t-1} \phi_t \tag{44}$$

$$K_t = \frac{P_{t-1}\phi_t}{\lambda_t + s_t} \tag{45}$$

$$\hat{\theta}_t = \hat{\theta}_{t-1} + K_t \epsilon_t \tag{46}$$

$$P_t = (I - K_t \phi_t^T) P_{t-1} \frac{1}{\lambda_t}$$
(47)

Stochastic Adaptive Control - Time-variant Estimation Time-varying systems - Fortescue's Method



If the variance is unknown, we can introduce an estimate

$$\lambda_{t} = 1 - \frac{1}{N_{0}} \frac{\epsilon_{t}^{2}}{r_{t}s_{t}}$$

$$r_{t} = r_{t-1} + \frac{1}{t} \left(\frac{\epsilon_{t}^{2}}{s_{t}} - r_{t-1} \right),$$

$$r_{0} = \epsilon_{0}^{2}$$
(48)
(48)

Stochastic Adaptive Control - Time-variant Estimation Time-varying systems - Model Estimators

Introduce model of parameters

$$\theta_{t+1} = \theta_t + v_t, \qquad v_t \sim N(0, R_1 \sigma^2)$$
(50)
$$y_t = \phi_t^T \theta_t + e_t, \qquad e_t \sim N(0, \sigma^2)$$
(51)

Estimate parameters using the Kalman filter

Data update

$$\hat{\theta}_{t|t} = \hat{\theta}_{t|t-1} + P_{t|t-1}\phi_t(y_t - \phi_t^T \hat{\theta}_{t|t-1})$$
(52)

$$P_{t|t}^{-1} = P_{t|t-1}^{-1} + \phi_t \phi_t^T$$
(53)

Time update

$$\hat{\theta}_{t+1|t} = \hat{\theta}_{t|t} \tag{54}$$

$$P_{t+1|t} = P_{t|t} + R_1 \tag{55}$$

Experiment design

When attempting to identify a system, consider the following:

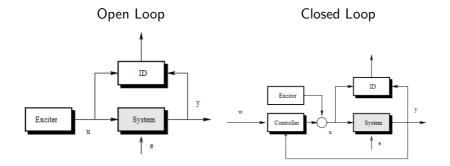
- **1** What are the outputs?
- **2** What are the inputs?
- **3** What are the disturbances?

Also consider some practical aspects of the system

- 1 What are we allowed to do?
- **2** What type of model are we interested in?

Stochastic Adaptive Control - Time-variant Estimation Design Configurations





For any system $\mathcal S,$ we can construct a set of models $\mathcal M$ to describe it

$$\mathcal{S}: \quad y = G_0(q)u + H_0(q)e \tag{56}$$

$$\mathcal{M} = \{ G(q, \theta), H(q, \theta) | \theta \in \mathcal{D} \}$$
(57)

Ideally, the system should be included in the set of possible models

$$S \in \mathcal{M}$$
 (58)

Given two models in \mathcal{M}

$$\mathcal{M}_1: y = G_1(q)u + H_1(q)e_1$$
(59)

$$\mathcal{M}_2: y = G_2(q)u + H_2(q)e_2$$
(60)

we want to be able to determine which that describes the system better

Therefore, we need to perform an informative (open-loop) experiment

Stochastic Adaptive Control - Time-variant Estimation Informative Experiments

We want to determine an input signal resulting in data that is sufficiently informative to dinstinguish between models in \mathcal{M}

For two models identified using data that is sufficiently informative, the expectation

$$\overline{\mathbb{E}}[\Delta\epsilon^2] = \lim_{N \to \infty} \frac{1}{N} \sum_{t=1}^N \mathbb{E}[\Delta\epsilon_t^2] = \int_{-\pi}^{\pi} \phi_1(w) + \phi_2(w) dw = 0$$
(61)

only holds if

$$\phi_2(w) = \left|\frac{H_0 \Delta H}{H_1 H_2}\right|^2 \sigma^2 = 0 \quad \Rightarrow \quad \Delta H(e^{jw}) \equiv 0 \tag{62}$$

$$\phi_1(w) = \left|\frac{1}{H_1}\right|^2 \left|\Delta G + \frac{G_0 - G_2}{H_2} \Delta H\right|^2 \Phi_u(w) = 0$$
(63)

$$\Rightarrow |\Delta G(e^{jw})|^2 \Phi_u(w) \equiv 0 \Rightarrow \Delta G(e^{jw}) \equiv 0$$
(64)

Consequently, the input should have a spectrum $\Phi_u(w)$ for which the above expectation only becomes zero for identical models in \mathcal{M} .

25 DTU Compute

Stochastic Adaptive Control - Time-variant Estimation Informative Experiments - Persistently excited signal

A quasi-stationary signal with spectrum $\Phi_u(w)$ is persistently excited of order n (pe(n)) if, for all filters in the form

$$M(q^{-1}) = m_0 + m_1 q^{-1} + \dots + m_{n-1} q^{-(n-1)}$$
(65)

the relation

$$\Phi_z(w) = |M(e^{jw})|^2 \Phi_u(w) = 0, \quad z_t = M(q^{-1})u_t$$
(66)

implies that for all w

$$M(e^{jw}) = 0 \tag{67}$$

 ${\cal M}(q^{-1})$ has n parameters and n-1 zeros; implying that ${\cal M}(q){\cal M}(q^{-1})$ has at most n-1 different zeros



Stochastic Adaptive Control 18.4.2024

Stochastic Adaptive Control - Time-variant Estimation Informative Experiments - Persistently excited signal

Equivalently, the spectrum, $\Phi_u(w),$ has to be non-zero at at least n different points in the interval $w\in [-\pi,\pi]$

The reason is that a signal which is pe(n) cannot be filtered to zero by an MA filter of order n-1, but n or higher might do it

$$u_t = \text{const} \neq 0$$
, signal is pe(1) (68)

$$M_1(q^{-1}) = 1 - q^{-1}$$
: $M_1(q^{-1})u_t = u_t - u_{t-1} = 0$ (69)

$$M_0(q^{-1}) = 1:$$
 $M_0(q^{-1})u_t = u_t \neq 0$ (70)

or looking at the spectrum: it is always zero

$$\Phi_u = \tilde{d}\delta(w) \tag{71}$$

$$\Phi_{M_1 u} = 2(1 - \cos(w))\tilde{d}\delta(w) = 0$$
(72)



Transfer function

$$G = q^{-k} \frac{B(q)}{A(q)} = q^{-k} \frac{b_0 + b_1 q^{-1} + \dots + b_{n_b} q^{-n_b}}{1 + a_1 q^{-1} + \dots + a_{n_a} q^{-n_a}}$$
(73)

The signal u_t has to be $pe(n_b + n_a + 1)$

$$\Delta G = \frac{B_1}{A_1} - \frac{B_2}{A_2} = \frac{B_1 A_2 - B_2 A_1}{A_1 A_2} = 0 \quad \Rightarrow \quad |B_1 A_2 - B_2 A_1|^2 \Phi_u(w) = 0$$
(74)

where it can be seen that the effective part of ΔG has the order $n_b + n_a$

Crest factor (for zero-mean signals)

$$C_r^2 = \frac{\max_t u_t^2}{\lim_{N \to \infty} \frac{1}{N} \sum_{t=1}^N u_t^2}$$
(75)

The crest factor should be as low as possible (the minimum is 1)

For binary signals, $u_t = \pm \bar{u}$, the crest factor is minimum, $C_r^2 = 1$

Consequently, binary signals are useful for linear systems, but cannot, in general, handle nonlinear systems

$$y_t = \frac{B(q)}{A(q)} f(u_t) \tag{76}$$

$$f(u_t) = \alpha \cos(\pm \bar{u}) = \alpha \cos(\bar{u}) \tag{77}$$

Single harmonic signal

$$u_t = A\sin(wt),\tag{78}$$

- Two non-zero frequency components in its spectrum (at $\pm w$)
- It is pe(2)
- Its crest factor is $C_r^2 = 2$

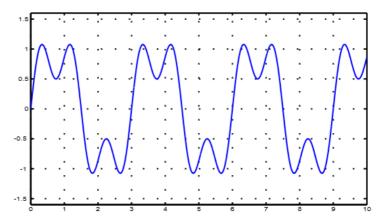
Sum of sines

$$u_t = \sum_{k=1}^{n} A_k \sin(w_k t + \phi_t)$$
(79)

- Two components for each w_k , so the signal is pe(2n)
- If $w_k = 0$ or $w_k = \frac{\pi}{T_s}$, the order goes down by 1 to pe(2n-1) (by 2 if both)
- The crest factor is, in the worst case, $C_r^2=2n,$ and lowest if the sinusoids are maximally out of phase

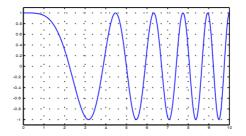


Sum of 2 harmonics, with maximum phase difference (180°)



Single sine function: The chirp signal

$$u_t = A\sin((w_0 + \alpha t)t), \quad C_r^2 = \sqrt{2}$$
 (80)

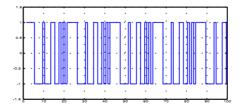


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PRBS signal

$$z_t = \text{mod}(B(q^{-1})z_{t-1}, 2)$$
(81)

B is order m and the signal has the maximum length $M=2^m-1$



• PRBS signals are deterministic, but have properties similar to those of white noise

• A PRBS signal is pe(M-1) and $C_r^2 = 1$

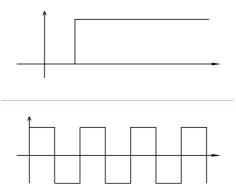


Alternative: Apply random Gaussian signals that are filtered/colored white noise signals

$$u_t = H_u(q)\check{e}_t, \quad \check{e}_t \sim \mathcal{F}_{iid}(0, \sigma_u^2)(white)$$
(82)

- In practice, we would have to use a truncated Gaussian to keep the control bounded, e.g., within $\pm 3\sigma$ ($\approx 99\%$ coverage), resulting in $C_r^2 = 3$
- Random binary signals can be generated by taking the sign of a suitable Random Gaussian signal

Step and square wave signals are also commonly used



For a step at time M and a square (both between d_0 and d_1)

$$C_r^2 = \frac{d_1^2}{\lim_{N \to \infty} \frac{M d_0^2 + (N - M) d_1^2}{N}} = \frac{d_1^2}{d_1^2 + \lim_{N \to \infty} \frac{M}{N} d_0^2} = 1, \qquad C_r^2 = \frac{d_1^2}{\frac{1}{2} d_1^2 + \frac{1}{2} d_0^2}$$

The pulse can also be represented as an infinite harmonic sum

35 DTU Compute

Stochastic Adaptive Control - Time-variant Estimation Informative experiments - Example



$$A(q^{-1})y_t = B(q^{-1})u_t + C(q^{-1})e_t, \qquad e_t \sim N(0, 0.05)$$
(83)

Polynomials

$$A(q^{-1}) = 1 - 1.2q^{-1} + 0.8q^{-2},$$
(84)

$$B(q^{-1}) = -0.5q^{-1} + 0.2q^{-2},$$
(85)

$$C(q^{-1}) = 1 + 0.3q^{-1} \tag{86}$$

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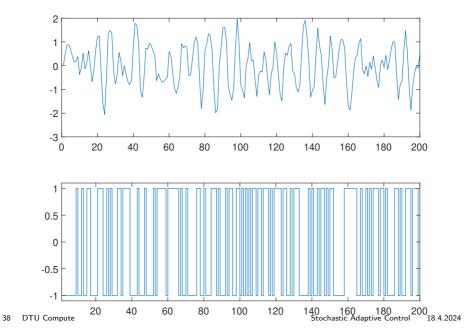
Define system

Simulate system

```
>> N = 200;
>> u = iddata([], idinput(N, 'prbs'));
>> e = iddata([], sqrt(R)*randn(N, 1));
>> y = sim(M, [u, e]);
>> simdata = [y.y, u.u];
```

Visualize simulation

```
>> figure(1);
>> subplot(211);
>> plot(y,y);
>> subplot(212);
>> stairs(u.u([1:end, end]));
>> xlim([1, numel(u.u)]);
>> ylim([-1.1, 1.1]);
```



Stochastic Adaptive Control - Time-variant Estimation Informative experiments - Exercise



$$A(q^{-1})y_t = B(q^{-1})u_t + C(q^{-1})e_t, \qquad e_t \sim N(0, 0.05)$$
(87)

Polynomials

$$A(q^{-1}) = 1 + 1.3q^{-1} - 0.7q^{-2},$$
(88)

$$B(q^{-1}) = -0.3q^{-1} - 0.1q^{-2},$$
(89)

$$C(q^{-1}) = 1 + 0.5q^{-1} \tag{90}$$

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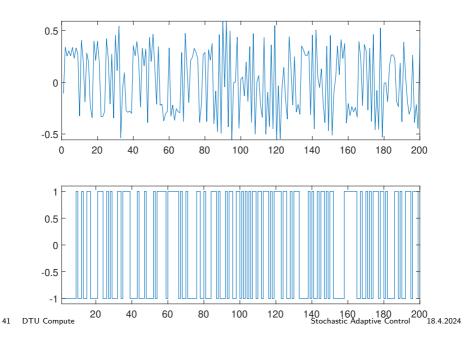
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Visualize simulation

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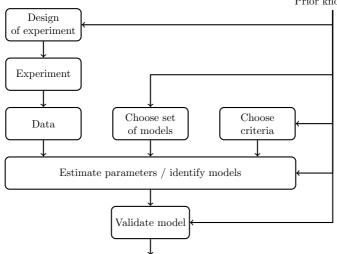
Model validation

We now know how to estimate a model, but how do we check if it is correctly estimated when we don't know the true parameters?

Essentially, we are asking the following two questions

- **1** Is our model too simple?
- **2** Is our model too complex?

Stochastic Adaptive Control - Time-variant Estimation Model Validations



Prior knowledge

Three available quantities for validation

- **1** the estimated parameters
- **2** the uncertainty (the variance)
- **③** the undescribed model parts (the residuals)

The last is the source of measurement deviations

$$measurement(y) = model(\theta, u) + residual(\epsilon)$$
(91)



Question: Does our model have too many parameters?

Unbiased estimate

$$\hat{\theta} \sim \mathcal{F}(\theta, P)$$
 (92)

 θ_i is significant if it, with reasonable certainty, is different from zero

Use a marginal parameter test to validate that a parameter is significant

For sufficiently many measurements, the distribution approaches a normal distribution

$$\hat{\theta} \sim N(\theta, P)$$
 (93)

If the following holds, θ_i is, with $(1 - \alpha)$ % confidence not insignificant

$$|\hat{\theta}_i| > f_{1-\frac{\alpha}{2}} \sqrt{P_{i,i}} \tag{94}$$

 f_x is the *x*th quantile of the standard normal distribution. This approach requires that the variance, *P*, is known

Stochastic Adaptive Control - Time-variant Estimation Model validation - parameter insignificant?

DTU

If the variance, P, was estimated, use the t-distribution

$$z_i = \frac{\hat{\theta}_i}{\sqrt{P_{i,i}}} \sim t(M - d_p) \tag{95}$$

 d_p is the number of parameters and M is the number of measurements

If the following holds, θ_i is, with $(1-\alpha)\%$ confidence not insignificant

$$|\hat{\theta}_i| > f_{1-\frac{\alpha}{2}}^t (M - d_p) \sqrt{P_{i,i}}$$
(96)

 f^t_x is the $x{\rm th}$ quantile of the t-distribution. Again, if $M\gg d_p,$ this will approach the normal distribution

(notheric of incignificant parameters $(\theta = 0)$

Test statistic for the hypothesis of insignificant parameters ($\theta_b = 0$)

$$z_b = \hat{\theta}_b^T P_b^{-1} \hat{\theta}_b \sim F(d_b, M - d_p)$$
(98)

If the following holds, all parameters in θ_b are, with $(1-\alpha)\%$ confidence significant

$$z_b > f_{1-\alpha}^F(d_b, M - d_p) \tag{99}$$

 $\begin{array}{ll} d_b \text{ is the size of the subset and } f_x^F \text{ is the } x \text{th quantile of the F-distribution.} \\ \text{For large } M \text{, we can apply a } \chi^2(d_b) \text{ instead of the F-distribution} \\ & \text{ Stochastic Adaptive Control} & \text{ 18.4.2024} \end{array}$

More than one parameter might be insignificant, but we cannot tell whether its some or all

 $\hat{\theta} = \begin{vmatrix} \hat{\theta}_a \\ \hat{\theta}_b \end{vmatrix} \sim N\left(\begin{vmatrix} \theta_a \\ \theta_b \end{vmatrix}, \begin{vmatrix} P_a & P_{ab} \\ P_{ab}^T & P_b \end{vmatrix} \right)$

But we can test whether all parameters in a subset θ_b are significant

(97)

Stochastic Adaptive Control - Time-variant Estimation Model Reduction

Distribution of parameter estimates

$$\begin{bmatrix} \theta_a \\ \theta_b \end{bmatrix} \sim N\left(\begin{bmatrix} \hat{\theta}_a \\ \hat{\theta}_b \end{bmatrix}, \begin{bmatrix} P_a & P_{ab} \\ P_{ab}^T & P_b \end{bmatrix} \right)$$
(100)

If a subset of the parameters, $\hat{\theta}_b$, is insignificant, we can reduce the model using the projection theorem

$$\theta_a | \theta_b \sim N(\hat{\bar{\theta}}_a, \bar{P}_a)$$
(101)

$$\hat{\bar{\theta}}_a = \hat{\theta}_a - P_{ab}^T P_b^{-1} \hat{\theta}_b \tag{102}$$

$$\bar{P}_a = P_a - P_{ab} P_b^{-1} P_{ab}^T$$
(103)

When have we used the projection theorem before and what for?

Think about it for yourself for <u>one minute</u> and then discuss with the person next to you for <u>one minute</u>.

Stochastic Adaptive Control - Time-variant Estimation Insignificant: singular analysis of the variance matrix *P*

Most estimations methods involve solving linear equations in the form

$$H\hat{\theta} = g \tag{104}$$

 ${\cal H}$ is a measure of the data set related to the variance, ${\cal H}^{-1}={\cal P}$

$$P = \left(\sum_{i=0}^{N} \psi_i \psi_i^T\right)^{-1} \sigma^2 \tag{105}$$

- If a model is overparameterized, then (in the ideal case) H will be singular
- \bullet In the less ideal case, H is invertible, but has eigenvalues that are significantly smaller than the rest

$$\operatorname{eig}(H)_i \ll \operatorname{eig}(H)_j \qquad \Leftrightarrow \qquad \operatorname{eig}(P)_i \gg \operatorname{eig}(P)_j$$
(106)

• This requires that the system is sufficiently excited – insufficiently excited systems will result in similar issues

Stochastic Adaptive Control - Time-variant Estimation Condition number

Another way to evaluate if a model is overparameterized is to consider the condition number of its variance.

$$\operatorname{cond}(P) = \frac{|\lambda_{\max}|}{|\lambda_{\min}|}, \quad \lambda = \operatorname{eig}(P)$$
 (107)

where λ_{\min} and λ_{\max} are the smallest and largest eigenvalues of P

If $\operatorname{cond}(P)$ is large, it indicates overparameterization

Example:

Model 1:
$$cond(P_1) = 1000$$
(108)Model 2: $cond(P_2) = 40$ (109)

Model 1 appears to be too complex, while model 2 is more balanced

Stochastic Adaptive Control - Time-variant Estimation Zeros and poles: Cancellation?

If the model is overparameterized, some zeros and poles might be close to each other

$$y_t = H_{yu}(q)u_t + H_{ye}(q)e_t$$
 (110)

Use linearization to approximate uncertainty in zero and poles

$$\hat{p}_i = f_i(\hat{\theta}) \simeq f_i(\theta) + \frac{\partial f_i}{\partial \theta} \tilde{\theta}, \qquad \tilde{\theta} \sim N(0, P)$$
 (111)

$$\hat{p}_i \sim N\left(p_i, \frac{\partial f_i}{\partial \theta} P\left(\frac{\partial f_i}{\partial \theta}\right)^T\right)$$
(112)

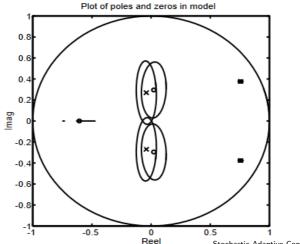
If the confidence intervals of a pole and a zero overlap, it is a strong indication that they cancel each other out

Hint: Use Matlab's zpplot

Stochastic Adaptive Control - Time-variant Estimation Zeros and poles: Example of cancellation

$$(1 - 1.5q^{-1} + 0.7q^{-2})y_t = (1 - 0.5q^{-1})u_t + e_t$$
(113)

$$(1 - a_1 q^{-1} + \dots + a_4 q^{-4})y_t = (b_0 + \dots + b_3 q^{-3})u_t + e_t$$
(114)



54 DTU Compute

Stochastic Adaptive Control 18.4.2024

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Stochastic Adaptive Control - Time-variant Estimation Residual Analysis

Question: Is the model too simple?

Residuals

$$measurement(y) = model(\theta, u) + residual(\epsilon)$$
(115)

For a perfect model, the residuals would have the following properties

- $\bullet \epsilon_t \sim \mathcal{F}(0, \sigma^2).$
- **2** ϵ_t has a symmetric distribution
- $\mathbf{8} \epsilon_t$ is white
- **(4** ϵ_t is uncorrelated with current and prior inputs

Equivalently (in terms of co-variance functions)

$$r_{\epsilon}(k) = \mathbb{E}[\epsilon_{t+k}\epsilon_t] = \begin{cases} \sigma^2 & k = 0, \\ 0 & \text{otherwise,} \end{cases} \qquad r_{\epsilon_t,u_t}(k) = \mathbb{E}[\epsilon_{t+k}u_t] = 0$$
(116)

Important: use one data set for *estimation* and another for the *validation* (cross-validation)

55 DTU Compute

Stochastic Adaptive Control - Time-variant Estimation Residual Analysis - mean and variance test



Simple approach: Test whether the distribution of the residuals has the right mean and variance

If the below holds, the residuals are not zero mean

$$|\bar{\epsilon}| > f_{1-\frac{\alpha}{2}}^t (M-1) \sqrt{\frac{S^2}{M}}$$
 (117)

$$\bar{\epsilon} = \frac{1}{M} \sum_{i=1}^{M} \epsilon_i, \quad S^2 = \frac{1}{M-1} \sum_{i=1}^{M} (\epsilon_i - \bar{\epsilon})^2$$
 (118)

If either of the below hold, the variance is time-varying

$$\frac{S_1^2}{S_2^2} < f_{\alpha/2}^F(M_1, M_2) \text{ or } \frac{S_1^2}{S_2^2} > f_{1-\alpha/2}^F(M_1, M_2)$$
(119)
$$S_i^2 = \frac{1}{M_i} \sum_{j=1}^{M_i} \epsilon_{i+j}^2$$
(120)

Note: The intervals must be non-overlapping

Test for whiteness: The number of sign changes, z, should follow (M is the number of data points)

$$z \sim N\left(\frac{M-1}{2}, \frac{M-1}{4}\right) \tag{121}$$

We reject the hypothesis if either of the below holds

$$z < \frac{M-1}{2} - \sqrt{\frac{M-1}{4}} f_{1-\frac{\alpha}{2}}^{N} \text{ or } z > \frac{M-1}{2} + \sqrt{\frac{M-1}{4}} f_{1-\frac{\alpha}{2}}^{N}$$
(122)

That is, the hypothesis is rejected if the test statistic is outside the confidence interval

Stochastic Adaptive Control - Time-variant Estimation Residual Analysis - test of co-variance function



Alternative test for whiteness: The auto-covariance must be in the form

$$r_{\epsilon}(k) = \mathbb{E}[\epsilon_{t+k}\epsilon_t] = \begin{cases} \sigma^2 & k = 0\\ 0 & \text{otherwise} \end{cases}$$
(123)

Estimates of auto-covariance and auto-correlation

$$\hat{r}_{\epsilon}(k) = \frac{1}{M} \sum_{t=1}^{M-k} \epsilon_{t+k} \epsilon_t, \qquad \hat{\rho}_{\epsilon}(k) = \frac{\hat{r}_{\epsilon}(k)}{\hat{r}_{\epsilon}(0)}$$
(124)

Test the covariance at each time step

$$H_0: \quad \sqrt{M}\hat{\rho}_{\epsilon}(k) \sim N(0,1), \text{ reject if } |\hat{\rho}_{\epsilon}(k)| > \frac{f_{1-\frac{\alpha}{2}}^N}{\sqrt{M}}$$
(125)

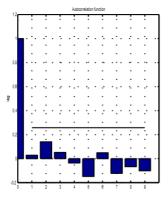
Test if the covariance is zero for $k \neq 0$

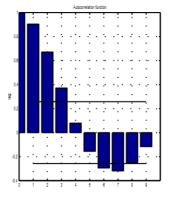
$$H_0: \quad z = M \sum_{i=1}^m \hat{\rho}_{\epsilon}^2(i) \sim \chi^2(m), \text{ reject if } z > f_{1-\alpha}^{\chi^2}(m)$$
 (126)

58 DTU Compute

Stochastic Adaptive Control - Time-variant Estimation Residual Analysis - test of autocorrelation







Stochastic Adaptive Control - Time-variant Estimation Residual Analysis - cross-covariance function test

Test the cross-covariance

$$r_{\epsilon,u}(k) = \mathbb{E}[\epsilon_{t+k}u_t] = 0 \tag{127}$$

Cross-covariance and cross-correlation

$$\hat{r}_{\epsilon,u}(k) = \frac{1}{M} \sum_{i=1}^{M-k} \epsilon_{t+k} u_t, \qquad \hat{\rho}_{\epsilon,u}(k) = \frac{\hat{r}_{\epsilon,u}(k)}{\sqrt{\hat{r}_{\epsilon}(0)\hat{r}_u(0)}}$$
(128)

Marginal test of the cross-covariance

$$H_0: \quad \sqrt{M}\hat{\rho}_{\epsilon,u}(k) \sim N(0,1), \text{ reject if } |\hat{\rho}_{\epsilon,u}(k)| > \frac{f_{1-\frac{\alpha}{2}}^N}{\sqrt{M}}$$
(129)

Check if the covariance is zero for $k\neq 0$

$$H_0: \quad z = M \sum_{i=1}^m \hat{\rho}_{\epsilon,u}^2(i) \sim \chi^2(m), \text{ reject if } z > f_{1-\alpha}^{\chi^2}(m)$$
 (130)

60 DTU Compute

Stochastic Adaptive Control - Time-variant Estimation Residual Analysis - spectral density test



Alternative test for whiteness: Consider the Fourier transformed residuals

$$X(w_k) = \frac{1}{M} \sum_{t=1}^{M} \epsilon_t e^{jw_k t}$$
(131)

Hint: Matlab's fft can be used to compute $X(w_k)$

Estimated spectral density (periodogram)

$$\hat{\phi}(w_k) = |X(w_k)|^2$$
 (132)

Hint: Matlab's effe can be used to compute $\hat{\phi}(w_k)$

If x_t is white noise

$$\mathbb{E}[\hat{\phi}(w_k)] = 2\sigma^2 \tag{133}$$

Stochastic Adaptive Control - Time-variant Estimation Residual Analysis - model comparison tests Question: Can we validate a model using a single data set?

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Coefficient of determination

$$R^{2} = \frac{J_{0} - J(\hat{\theta})}{J_{0}}$$
(134)
$$J_{0} = \frac{1}{2} \sum_{i=1}^{M} (y_{i} - \bar{y})^{2}, \quad J(\hat{\theta}) = \frac{1}{2} \sum_{i=1}^{M} \epsilon_{i}^{2}$$
(135)

 $J(\hat{\theta})$ is the loss-function and a perfect model results in $R^2=1.$ Lower values of R^2 indicate worse models

Alternative loss functions

$$W(\hat{\theta}) = \sum_{i=1}^{M} \epsilon_i^2, \qquad W_M(\hat{\theta}) = \frac{1}{M} \sum_{i=1}^{M} \epsilon_i^2$$
(136)

 The loss functions are monotonically decreasing with model complexity

 62
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 Stochastic Adaptive Control
 18.4.2024

Stochastic Adaptive Control - Time-variant Estimation Residual Analysis - model comparison: F-test



Objective: Compare two model classes, \mathcal{M}_1 and \mathcal{M}_2 using the F-test

Hypothesis: $\mathcal{M}_{true} \subset \mathcal{M}_1 \subset \mathcal{M}_2$ where $d_2 \geq d_1$ are the number of model parameters. Consequently, the loss-function $J_i = J_i(\hat{\theta})$ does not decrease significantly by increasing the model size if $\mathcal{M}_1 \subset \mathcal{M}_2$

Test statistic

$$H_0: z = \frac{J_1 - J_2}{J_2} \frac{M - d_2}{d_2 - d_1} \sim F(d_2 - d_1, M - d_2)$$
(137)

Reject hypothesis if

$$z > f_{1-\alpha}^F(d_2 - d_1, M - d_2)$$
(138)

Stochastic Adaptive Control - Time-variant Estimation Residual Analysis - model comparison: Information Criteria Information criteria

1 Akaike's Information Criterion (AIC); tends towards higher complexity

$$AIC = \left(1 + \frac{2d}{M}\right) W_M \tag{139}$$

2 Bayesian Information Criterion (BIC);

$$BIC = \left(1 + \frac{\log(M)d}{M}\right) W_M \tag{140}$$

3 Akaike's Final Prediction Error (FPE) Criterion; expresses the variance of the prediction error, also $FPE \rightarrow AIC, M \gg d$

$$FPE = \frac{M+d}{M-d}W_M = \left(1 + \frac{2d}{M-d}\right)W_M \tag{141}$$

 If two models have the same d, choose the one with the lowest loss function

 64
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 18.4.2024



$$BIC = \left(1 + \frac{\log(M)a}{M}\right) W_M \tag{143}$$

$$FPE = \frac{M+d}{M-d}W_M = \left(1 + \frac{2d}{M-d}\right)W_M \tag{144}$$

What happens to the criteria as $M \to \infty$?

Think about it for yourself for one minute and then discuss with the person next to you for one minute. (142)

Information criteria

Stochastic Adaptive Control - Time-variant Estimation Model validation - Example

ARMAX model

$$A(q^{-1})y_t = B(q^{-1})u_t + C(q^{-1})e_t, \qquad e_t \sim N(0, 0.05)$$
(145)

Polynomials

$$A(q^{-1}) = 1 - 1.2q^{-1} + 0.8q^{-2},$$
(146)

$$B(q^{-1}) = -0.5q^{-1} + 0.2q^{-2},$$
(147)

$$C(q^{-1}) = 1 + 0.3q^{-1} \tag{148}$$

Define system

Simulate system

```
>> N = 200;
>> u = iddata([], idinput(N, 'prbs'));
>> e = iddata([], sqrt(R)*randn(N, 1));
>> y = sim(M, [u, e]);
>> simdata = [y.y, u.u];
```

Stochastic Adaptive Control - Time-variant Estimation

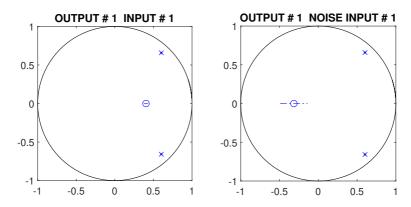
Model validation - Example

Estimate parameters

>> th = armax(simdata, [2, 2, 1, 1])

Zero-pole cancellation





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Eigenvalue analysis

Insignificant parameters

Stochastic Adaptive Control - Time-variant Estimation Model validation - Exercise ARMAX model

$$A(q^{-1})y_t = B(q^{-1})u_t + C(q^{-1})e_t, \qquad e_t \sim N(0, 0.05)$$
(149)

Polynomials

$$A(q^{-1}) = 1 - 1.2q^{-1} + 0.8q^{-2},$$
(150)

$$B(q^{-1}) = -0.5q^{-1} + 0.2q^{-2},$$
(151)

$$C(q^{-1}) = 1 + 0.3q^{-1} \tag{152}$$

Estimate the parameters in a model with

$$n_a = 3, \tag{153}$$

$$n_b = 3, \tag{154}$$

$$n_c = 2, \tag{155}$$

$$k = 1 \tag{156}$$

Solve the exercise in 15 min.

Stochastic Adaptive Control - Time-variant Estimation

Model validation - Exercise

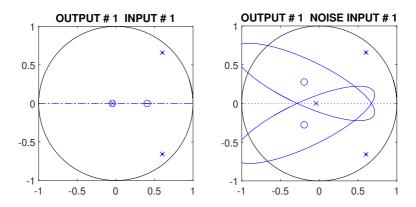
Estimate parameters

>> th = armax(simdata, [3, 3, 2, 1])

Zero-pole cancellation



>> zpplot(th2zp(th, 0), norminv(0.995)) % 99% Gaussian Cl



Eigenvalue analysis

Insignificant parameters

>> TH + sqrt(diag(P)) * [-1,1] * norminv(0.995)ans = -2.99200.6649 -1.45832.9499 -1.42321.4888 -0.5054-0.4884-0.72911.0882 -0.35710.3749 -1.42562.2050 -0.47900.7077

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Stochastic Adaptive Control - Time-variant Estimation Questions



Questions?