### **State Exploration for Real-Time**

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# What you'll learn

- Alur-Dill timed automata:
  - The model
  - use in verification
  - finitariness: clock regions
- Clock zones as a symbolic representation for TA states:
  - represent (certain) convex unions of clock regions, avoiding the exponential blowup of the region construction
- Difference bound matrices (DBMs)
  - a practical representation of clock zones

#### **Timed Transition Systems**



- Stays in state opening for exactly 6 time units,
- stays in state open between 4 and 8 time units,
- stays in state closing for exactly 6 time units,
- stays in state closed for exactly 5 time units,
- stays in all other states arbitrarily long.
- ⇒ Parallel composition is not transition-synchronous!

# Formal setup

A timed transition system TTS = (V, E, L, T,  $\alpha$ , G, R, Inv, I) over a set C of clocks and alphabet  $\Sigma$  has

- a set V of vertices (interpreted as discrete system states, a.k.a. locations),
- a set E of edges (interpreted as possible transitions),
- $L \in V \rightarrow \mathcal{P}(AP)$  labels the vertices with atomic propositions that apply in the individual vertices,
- $I \subseteq V$  is a set of initial states,
- $T: E \rightarrow (V \times V)$  maps edges to location changes,
- $\alpha : E \to \Sigma$  assigns a communication to transitions,
- $G: E \to \mathcal{P}(ClockVal)$  gives conditions for a transition to be taken,
- $R: E \to \mathcal{P}(C)$  states the clocks to be reset upon a transition,
- Inv:  $V \to \mathcal{P}(ClockVal)$  yields state invariants denoting when a state may be held,

where  $ClockVal = C \rightarrow \mathbb{R}_{>0}$ .

# **Runs of TTS**

Given a TTS (V, E, L, T,  $\alpha$ , G, R, In $\nu$ , I), a run r of the TTS is

- an alternating sequence  $(v_0, c_0) \stackrel{(e_0, t_0)}{\rightarrow} (v_1, c_1) \stackrel{(e_1, t_1)}{\rightarrow} \dots$  of
  - 1. state/clock-valuation pairs  $(v_i, c_i) \in V \times ClockVal$ ,
  - 2. transition/time pairs  $(e_i, t_i) \in E \times \mathbb{R}_{\geq 0}$
- with non-decreasing time stamps:  $t_i \leq t_{i+1}$  for each i
- that starts in an initial state:  $v_0 \in I$  and  $c_0 \equiv 0$
- and is state-transition-consistent:  $T(e_i) = (v_i, v_{i+1})$  for each i
- and satisfies the transition guards:  $c_i + (t_i t_{i-1}) \in G(e_i)$  for each i, where c + t(x) = c(x) + t for each clock x and  $t_{-1} = 0$ ,
- and invariably satisfies the state invariants:  $c_i + t \in Inv(v_i)$  for each i and each t with  $0 \le t \le t_i t_{i-1}$

• and obeys clock resets: 
$$c_{i+1}(x) = \begin{cases} c_i(x) + (t_i - t_{i-1}) & \text{iff } x \notin R(e_i) \\ 0 & \text{iff } x \in R(e_i) \end{cases}$$

for each i and each clock x.

# The quest

- The set of states of a TTS is  $V \times ClockVal$ .
- It is infinite, as  $ClockVal = C \rightarrow \mathbb{R}_{\geq 0}$ .
- Naive forward or backward (on the fly or symbolic) state coloring algorithms need not terminate.

#### Is reachability analysis etc. nevertheless mechanizable?

## **Simple clock constraints**

A clock constraint is simple iff

- it is of the form x ~ k, where x is a clock, k an integer constant, and ~ one of <, ≤, =, ≥, >
- a conjunction of such simple constraints.

From now on, we will concentrate on TTS where

- all guards are simple,
- all invariants are simple.

## **Clock regions**



### **Time-abstract bisimulation**

A time-abstract bisimulation between two TTS is a relation

 $\sim \subset (V \times ClockVal) \times (V' \times ClockVal')$ 

s.t. for each  $(\nu,c)\sim (\nu',c')$ :

1. if there is  $(v_1, c_1) \in V \times ClockVal$  and  $(e, t) \in E \times \mathbb{R}_{\geq 0}$  s.t.

$$(v,c) \stackrel{(e,t)}{\rightarrow} (v_1,c_1)$$

then there is  $(v'_1, c'_1) \in V' \times ClockVal'$  and  $(e', t') \in E' \times \mathbb{R}_{\geq 0}$  s.t.

$$(\nu',c') \stackrel{(e',t')}{\to} (\nu'_1,c'_1) \quad \text{and} \quad \alpha(e) = \alpha(e') \quad \text{and} \quad (\nu_1,c_1) \sim (\nu'_1,c'_1)$$

#### N.B.: t and t' are not related! $\rightarrow$ time abstraction.

#### **Time-abstract bisimulation (cntd.)**

2. if there is  $(v'_1, c'_1) \in V' \times ClockVal'$  and  $(e', t') \in E' \times \mathbb{R}_{\geq 0}$  s.t.

$$(\nu',c') \stackrel{(e',t')}{\rightarrow} (\nu'_1,c'_1)$$

then there is  $(v_1, c_1) \in V \times ClockVal$  and  $(e, t) \in E' \times \mathbb{R}_{\geq 0}$  s.t.

 $(\nu,c) \stackrel{(e,t)}{\to} (\nu_1,c_1) \quad \text{and} \quad \alpha(e) = \alpha(e') \quad \text{and} \quad (\nu_1,c_1) \sim (\nu_1',c_1')$ 



States in the  $\sim$  relation follow similar (same labels, different timing) traces.

## **Clock regions vs. time-abstract bisimulation**

**Thm.:** If ~ is a time-abstract bisimulation *on a TTS* s.t. ~ does only relate identical vertices (yet with potentially different clock val.s) and if  $(v, c) \sim (v', c')$  then a vertice  $w \in V$  is reachable from (v, c) iff w is reachable from (v', c').

**Thm.:** For any TTS, the relation  $\sim \subset (V \times ClockVal) \times (V \times ClockVal)$  defined by  $(\nu, c) \sim (\nu', c')$  iff

1. v = v',

- 2. F.e. clock x,  $\lfloor c(x) \rfloor = \lfloor c'(x) \rfloor$  or c(x) > mc < c'(x),
- 3. F.e. clock x, fract(c(x)) = 0  $\iff$  fract(c'(x)) = 0 or c(x) > mc < c'(x),
- 4. F.e. clock x, y, fract(c(x))  $\leq$  fract(c(y))  $\iff$  fract(c'(x))  $\leq$  fract(c'(y)) or ...

is a time-abstract bisimulation on the TTS (i.e., between the states of just that one TTS).

(mc is the maximum time constant in the TTS.)

**Cor.:** Wrt. vertex reachability (and other time-abstract notions like existence of time-abstract traces), states in the above  $\sim$  relation are indistinguishable.

**Obs.:** For any TTS, there are only finitely many equivalence classes wrt.  $\sim$ .

### Equivalence classes of $\sim$



## The region automaton

Given the TTS = (V, E, L, T,  $\alpha$ , G, R, In $\nu$ , I), we define its region "automaton" (like the TTS, it actually lacks an acceptance condition) to be the finite Kripke structure  $A_{TTS} = ([V \times ClockVal]_{\sim}, \rightarrow, L', [I \times \{x \mapsto 0\}]_{\sim})$  with

- $x \rightarrow y$  iff there is  $(v, c) \in x$ ,  $(v', c') \in y$ ,  $t \ge 0$ , and  $e \in E$  s.t.  $(v, c) \stackrel{(e,t)}{\rightarrow} (v', c')$
- L'([v, c]) = L(v).
  - This is a finite Kripke structure that can be subjected to CTL model-checking etc.
  - but its size is exponential in the number of clocks: #regions =  $|C|! \cdot 2^{|C|} \cdot \prod_{c \in C} (2 \max(c) + 2)$
  - Can we do the state-space traversal more symbolicly, representing sets of regions by predicates?

#### **Clock zones**

## **Clock zones**

A clock zone is the set of satisfying assignments in  $\mathbb{R}^n_{\geq 0}$  of a conjunction of

- inequations that compare a clock to an integer constant and
- inequations that compare the difference of two clocks to an integer constant.

By introduction of a dedicated clock  $x_0$  representing the value 0, difference logic formulae of the specific conjunctive form

$$\begin{split} \varphi & ::= \bigwedge_{x \in C} (x_0 - x \le 0) \land \bigwedge_{i=1}^n \psi_i \\ \psi_i & ::= c_{i1} - c_{i2} \sim_i k_i \\ \sim_i & ::= < | \le \\ k_i & ::\in \mathbb{Z} \end{split}$$

form an appropriate symbolic representation of clock zones.

## **Closure properties of clock zones**

If  $\varphi$  and  $\psi$  are symbolic representations of clock zones and  $d\in\mathbb{N}$  then symbolic representations

- $\phi \land \psi$  for zone intersection:  $\llbracket \phi \land \psi \rrbracket \stackrel{\text{def}}{=} \{ \vec{x} \in \mathbb{R}^n_{\geq 0} \mid \vec{x} \models \phi \text{ and } \vec{x} \models \psi \}$
- $\exists x_i . \phi$  for clock hiding:

$$\llbracket \exists x_{i}.\varphi \rrbracket \stackrel{\text{def}}{=} \left\{ (x_{1},\ldots,x_{n}) \middle| \begin{array}{l} \text{there is } y \in \mathbb{R}_{\geq 0} \text{ s.t.} \\ (x_{1},\ldots,x_{i-1},y,x_{i+1},\ldots,x_{n}) \models \varphi \end{array} \right\}$$

- $\phi[x_i := 0]$  for clock reset:  $\llbracket \phi[x_i := 0] \rrbracket \stackrel{\text{def}}{=} \llbracket x_i = 0 \land \exists x_i. \phi \rrbracket$
- $\phi \uparrow$  for elapse of time:

$$\llbracket \phi \uparrow \rrbracket \stackrel{\text{def}}{=} \{ (x_1 + \delta, \dots, x_n + \delta) \, | \, (x_1, \dots, x_n) \models \phi, \delta \in \mathbb{R}_{\geq 0} \, \}$$

can be obtained effectively.

# TA reachability using zones: the idea

- 1. Represent reachable state sets by lists of pairs of locations and clock zones  $\langle (l_1, z_1), \dots, (l_m, z_m) \rangle$ ,
- 2. for such a pair, compute the set  $Post_t(l, z)$  of successors under a transition t with T(t) = (l, l') by
  - let time elapse starting from z: φ<sub>1</sub> = z ↑ represents states reachable under arbitrary passage of time
  - intersect φ₁ with Inv(l): φ₂ = φ₁ ∧ Inv(l) reflects states reachable through time passage consistent with the location invariant (N.B.: invariant is convex due to simplicity)
  - intersect  $\phi_2$  with guard G(t):  $\phi_3 = \phi_2 \wedge G(t)$  reflects states reachable through time passage which enable the transition t
  - reset the clocks in R(t):  $\varphi_4=\varphi_3[r_1:=0]\dots[r_j:=0],$  where  $\{r_1,\dots,r_j\}=R(t),$  reflects the clock readings after performing t's resets
  - intersect with the target loc.'s invariant:  $\phi_5 = \phi_4 \wedge Inv(l')$
  - do the location change:  $Post_t(l, z) = (l', \phi_5)$ .

#### The state-space exploration

- 1. Start with the state list  $R_0 = I \times \{ `` \bigwedge_{x \in C} (x_0 x \le 0) \land \bigwedge_{x \in C} (x x_0 \le 0)'' \}.$
- 2. Repeat
  - (a) select  $(l_i, z_i) \in R_k$  and  $t \in E$  with source  $l_i$  s.t.  $Post_t(l_i, z_i)$  is not already subsumed by  $R_k$ ,
  - (b) build  $R_{k+1} = R_k \cdot \langle \text{Post}_t(l_i, z_i) \rangle$
  - until no such  $(l_i, z_i) \in R_k$  and  $t \in E$  is found.

#### **N.B.** Subsumption test can be performed at various levels of detail.

# The problem

 Iterating Post<sub>t</sub>(1, z) for all pairs (1, z) in the list of reachable states and all transitions need not terminate:



- In the region graph, we solved the problem by not distinguishing clock readings above the max. clock constant.
- We can achieve a similar effect by widening zones that extend beyond the max. clock constant:
  - Any constraint of the form  $x_i x_j \sim l$  with l > maxconstant is removed from the symbolic representation when it arises.

#### **Difference Bound Matrices**

## **Difference Bound Matrices**

Difference bound matrices (DBMs) are a canonizable representation for *conjunctive* formulae in difference logic

$$\begin{split} \varphi & ::= \bigwedge_{i=1}^{n} \psi_{i} \\ \psi_{i} & ::= c_{i1} - c_{i2} \sim_{i} k_{i} \\ \sim_{i} & ::= < | \leq \\ k_{i} & ::\in \mathbb{Z} \end{split}$$

Given a finite clock set C (in practice containing the pseudo-clock  $x_0$ ), a DBM M over C is a mapping

$$\underbrace{(C \times C)}_{\text{clock pairs}} \to (\underbrace{\{<,\le\} \times \mathbb{Z}}_{\text{constraint on diff.}} \cup \underbrace{\{(<,\infty)\}}_{\text{unconstrained}})$$

Encoding:  $M(x, y) = (\sim, k) \triangleq x - y \sim k$ 

## Implied constraints and tightening

**Observation:**  $x - y \sim_1 k_1$  and  $y - z \sim_2 k_2$  implies  $x - z \sim k_1 + k_2$ , where

$$\sim = \begin{cases} \sim_1 & \text{iff } \sim_1 = \sim_2 \\ < & \text{otherwise.} \end{cases}$$

Consequence: A DBM may contain constraint pairs which imply constraints that are tighter than the recorded constraints:  $M(x,y) = (\sim_1, k_1) \land M(y,z) = (\sim_2, k_2) \land M(x,z) = (\sim, k)$  and 1.  $k > k_1 + k_2$  or 2.  $k = k_1 + k_2$  but  $\sim = <$ , yet  $\sim_1 = <$  or  $\sim_2 = <$ .

**Solution:** *Tighten the DBM* by replacing the constraint by the stronger implied constraint.

Repeat this until no implied constraint stronger than a recorded constraint remains. This brings the DBM into a *canonical form*. Such canonization of DBMs can be done in cubic time using the *Floyd-Warshall algorithm*.

## **Properties of canonical DBMs**

Thm: A canonical DBM is unsatisfiable iff there is some  $x \in C$  such that M(x, x) = (<, 0) or  $M(x, x) = (\sim, k)$  with k < 0.

**Cor:** Satisfiability test of *canonical* DBMs runs in O(|C|) time.

## **Operations on clock zones using ca. DBMs**

Intersection:

$$M \wedge N(x, y) = \begin{cases} M(x, y) & \text{if } M(x, y) \text{ is tighter than } N(x, y) \\ N(x, y) & \text{otherwise} \end{cases}$$

**Clock reset:** When the dedicated clock variable  $x_0$  is used,

$$M[z := 0](x, y) = \begin{cases} M(x, y) & \text{if } x \neq z \text{ and } y \neq z \\ M(x, x_0) & \text{if } x \neq z \text{ and } y = z \\ M(x_0, y) & \text{if } x = z \text{ and } y \neq z \\ (\leq, 0) & \text{if } x = y = z \end{cases}$$

Note that canonicity saves an explicit quantifier elimination as the implied constraints are already in place!

#### These operations do not preserve canonicity!

### **Operations on clock zones using can. DBMs**

**Elapse of time:** When the dedicated clock variable  $x_0$  is used,

$$M \uparrow (x,y) = \begin{cases} M(x,y) & \text{if } x = x_0 \text{ or } y \neq x_0 \\ (<,\infty) & \text{if } x \neq x_0 \text{ and } y = x_0 \end{cases}$$

Widening: When the maximum clock constant is k,

$$\widetilde{M}(x,y) = \begin{cases} M(x,y) & \text{if } M(x,y) = (\sim, l) \text{ with } |l| \le |k| \\ (<, \infty) & \text{otherwise} \end{cases}$$

## **Pros and cons**

- Zone-based reachability analysis usually is explicit wrt. discrete locations:
  - maintains a list of location-zone pairs or
  - maintains a list of location-DBM pairs
  - confined wrt. size of discrete state space
  - avoids blowup by number of clocks and size of clock constraints through symbolic representation of clocks
- Region-based analysis provides a finite-state abstraction, amenable to finite-state symbolic MC
  - less dependent on size of discrete state space
  - exponential in number of clocks